

Disturbance Observer based Friction Compensator for a DC Motor

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Abstract– Friction is primary disturbance in motion control. Different types of friction cause diminution of original torque in a DC motor, such as static friction, viscous friction etc. By some means if those can be determined and compensated, the friction effect from the DC motor can be neglected. It would be a great advantage for control systems. Authors have determined the types of frictions as well as frictional coefficients and suggested a unique way of compensating the friction in a DC motor using Disturbance Observer Method which is used to determine the disturbance torques acting on a DC motor. In simulation approach, the method is modelled using MATLAB and the results have been obtained and analysed. The block diagram consists with DC motor model with DOB and RTOB. Practical approach of the implemented block diagram is shown by the obtained results. It is discussed the possibility of applying this to real life applications.

Keywords: Disturbance observer, Friction compensation

I. INTRODUCTION

In the field of automation, two communal problems which are faced by researchers and engineers are variation of friction forces and variation of inertia. As far as friction term is concerned, by definition the “Friction” term stands for the resistance to relative motion between two contacting bodies [1]. Since friction dictates terms in motion control applications which will influence to the design criteria and hardware configurations especially to results, it is imperative to ponder the possibility of neutralizing friction forces somehow. Typically, the processes of non-linear friction forces are unavoidable in high performance motion control systems. One simple solution would be to use high quality mechanical components such as DC Motors. But those are costly which escalates the cost of the application way beyond the desired limit and yet the effect is not disinterested comprehensively. Another method would be to use appropriate friction compensation method with relevant computing technology and combined with good mechanical components[2]. This will remove the friction effect. The core of this research is to provide an appropriate friction compensation method for a particular DC Motor and analyzing the outcome.

Here in this research, Friction compensation is based on a reasonable accurate model for nonlinear friction which can be obtained by off-line experiment. However, It is very hard to compensate for some friction components since those tend to change when the velocity, system overall load and configuration change. So the most viable thing to do is find

the friction coefficients initially using off-line experiment. Normally, those coefficients don’t change in a big tolerance. Nevertheless, there is a slight temperature and humidity effect for coefficients which can be neglected.

The aim of the research is to provide a new scheme to effectively compensate friction for a DC motor with guaranteed high precision and smoothness of motion. In this paper, Disturbance Observer (DOB) as well as Reaction Torque Observer (RTOB) prior to PID is used to compensate the frictional effects.

II. DC MOTOR MODELING

DC motor is the modest machine which converts electrical energy to the mechanical energy. Electrical equivalent model of a DC motor is shown in Fig.1

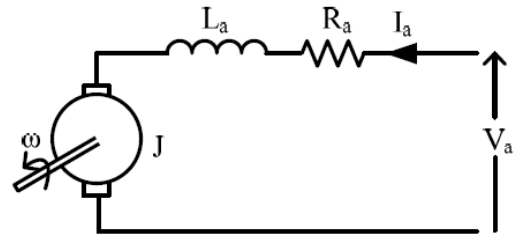


Fig.1. Electrical model of a DC Motor

A major fraction of the torque generated in the rotor of the motor is available to drive an external load. DC motors are widely used in numerous control applications, including robot manipulators, disk drive, machine tools and servo actuators and high torque application etc. DC motor is very popular especially in precise motion control applications, as the modeling is much simpler than all the other motors, Fig.1 shows the electrical model of a DC motor where R_a and L_a represents the armature resistance and inductance respectively[3].

Applying Kirchhoff’s voltage law to the DC motor model, following differential equation (1) can be obtained.

$$V_a = L_a \frac{dI_a}{dt} + R_a I_a + E_b \quad (1)$$

Where E_b is the back emf (2).

$$E_b = K_e \omega(t) \quad (2)$$

Where K_e is the back emf constant, then equation (1) becomes equation no. (3)

$$V_a = L_a \frac{dI_a}{dt} + R_a I_a + K_e \omega(t) \quad (3)$$

If the total mechanical output given by the motor is T_m , I_a can be written as equation no.(4)

$$I_a = \frac{T_m}{K_t} \quad (4)$$

Where K_t is the torque constant of the motor.

Applying (4) in (3) gives (5),

$$V_a = L_a \frac{d(T_m / K_t)}{dt} + R_a \frac{T_m}{K_t} + K_e \omega(t) \quad (5)$$

Motor torque T_m can be written as follows (6),

$$T_m = J \frac{d\omega}{dt} + B\omega + T_l + T_f \quad (6)$$

Where,

- T_f : Static Friction
- T_l : Load Torque
- T_m : Motor Torque
- J : Inertia Constant
- B : Viscosity coefficient
- ω : Rotor Speed

Equation (5) can be written in Laplace domain as (7) and Equation (4) can be written in Laplace domain as (9)

$$\omega(s) = \frac{T(s) - [T_l(s) + T_f(s)]}{Js + B} \quad (7)$$

$$T(s) = K_t I_a(s) \quad (8)$$

III. MOTOR FRICTION

Friction is the tangential reaction force between two contact surfaces. In dry sliding contacts between flat surfaces, friction can be modeled as elastic and plastic deformation forces of microscopical asperities. In dry rolling contact, friction is the result of a non-symmetric pressure distribution in the contacts. When a lubricant is added, other physical variations also will appear to the total system. For low velocities, the lubricant acts as a surface film, where the shear strength determines the friction. At higher velocities and low pressures a fluid layer of lubricant is built up in the surface due to hydrodynamic effects the behavior of friction has been extensively examined during the 20th century. The experiments have been performed under idealized conditions with clean surfaces and for stationary conditions, e.g. constant velocity. Friction is a linear combination of Coulomb friction, stiction, viscous friction, and stibeck effect[4].

A. Static Friction

Static friction is the friction when the two surfaces are sticking. The force required to overcome the static friction and initiate motion is called the breakaway force. The static friction is usually larger than Coulomb friction force.

B. Coulomb Friction

Coulomb friction is independent of velocity and is always present. This friction component is dependent only on the direction of motion, in such way that it is in the direction opposite to the velocity. The magnitude of Coulomb friction depends on the properties of the surfaces in contact and the normal force. Coulomb friction is also known as kinetic friction.

C. Viscous Friction

Viscous friction depends on the velocity. At zero velocity the viscous friction is zero and the viscous component increases with the increase of velocity

IV. DETERMINING FRICTION COEFFICIENTS

To estimate the friction components, constant angular velocity motion test is performed. Here disturbance observer is used as a torque sensor, which identifies all parameters included in the motion equation by using the estimated disturbance torque of the observer. Under the constant angular velocity motion test, the disturbance observer of a DC motor estimates the friction effect.

The disturbance observer structure is given as in equation (9). In the constant angular velocity motion test, the velocity is constant and the acceleration is zero, thus there is no effect of inertia. The torque constant coefficient calculated is equal to the nominal value and since the test is conducted under the no-load condition, there is no external torque affecting to the system and T_{int} . Interactive torque is assumed to be negligible effect on the system. Now the observer structure will be,

$$T_{dis} = B\dot{\theta} + T_f \quad (9)$$

At first to calculate the friction components of the DC motor the test was conducted with different velocities ranging from +1500 rpm to -1500 rpm which is shown in Fig.2. To obtain the better velocity response, the system used is tuned by velocity controller with disturbance observer. The disturbance torque was measured by the disturbance observer for different velocities.

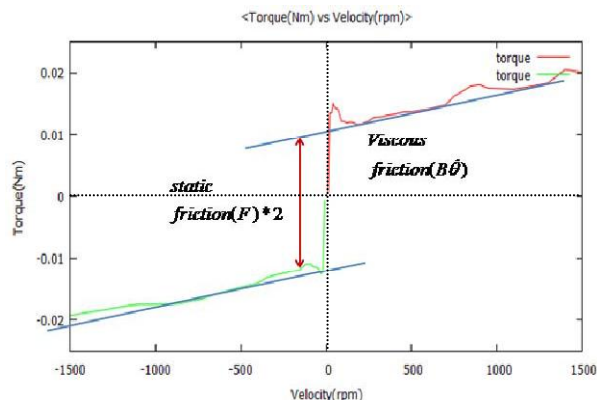


Fig.2. Friction variation with Speed in a DC Motor

V. FRICTION COMPENSATOR MODELING

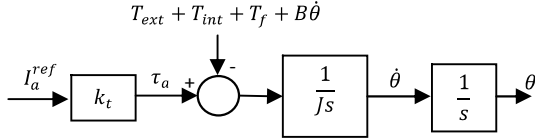


Fig.3. Disturbance Torque

Fig. 3 shows the disturbance torque block. Here I_a^{ref} is the torque current command determined by the DC motor motion controller. K_t is the torque constant which may vary according to the rotor position. The variation of the torque constant due to the space harmonics of the flux distribution generates the motor torque pulsation. τ_a is the generated motor torque obtained by the product of I_a^{ref} and K_t . The actual inertia constant J changes from the nominal inertia constant J_n , according to the altitude, the position of the rotor or after manufacturing modifications. The load torque includes T_{int} , the interactive torque which is defined in this research to have the elements of the pay load term, the coriolis term including the centrifugal term, and the coupled inertia term. Obviously T_{int} is defined not to include the self-inertia variation torque. T_{ext} is the reaction torque when mechanical system does the torque task. The friction has the components of both the static friction and the viscosity. T_f is the static friction torque and B is the viscosity coefficient. The disturbance torque is defined to have all the above components[5].

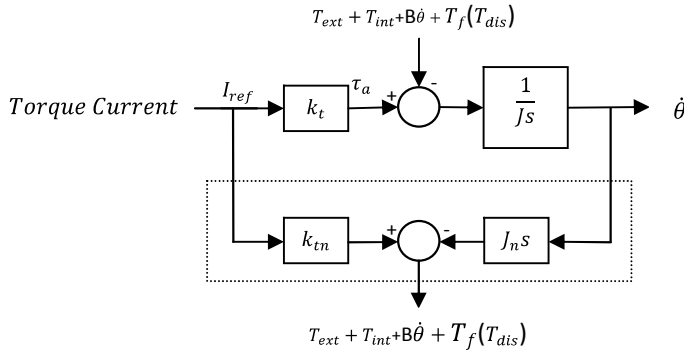


Fig.4. Determination of Disturbance Torque

It is difficult to realize Fig.4 directly, because the acceleration is not detected by ordinary sensors. Instead of that, the acceleration signal is calculated through the differentiation of the velocity of the encoder. In this calculation process, a low pass filter should be inserted to suppress high frequency noise. So that's the reason why the

block diagram in Fig.5 has been used for the further modifications.

In the robust control, it is essential to suppress the effect of the disturbance torque. For this purpose, the observer is introduced to identify it. From the above discussion the disturbance torque is defined as equation (10) to include the nonlinear effect of the load torque[6][7].

$$T_{dis} = T_{int} + T_{ext} + (J - J_n)\ddot{\theta} + B\dot{\theta} + T_f + (K_{tn} - K_t)I_a^{ref} \quad (10)$$

Here,

- T_{int} : Interactive torque
- T_{ext} : Reactive torque in torque task
- f : Static Friction
- I_a^{ref} : Torque current
- J : Inertia constant
- J_n : Nominal Inertia constant
- B : Viscosity coefficient
- k_t : Torque constant
- g_{rec} : RTOB filter coefficient
- k_{tn} : Nominal Torque constant
- $\dot{\theta}$: Angular velocity
- s : Laplace operator

$$\Delta J = J_n - J \quad (11)$$

$$\Delta K_t = K_{tn} - K_t \quad (12)$$

Equation (11) and Equation (12) represent variation of inertia and motor constant respectively.

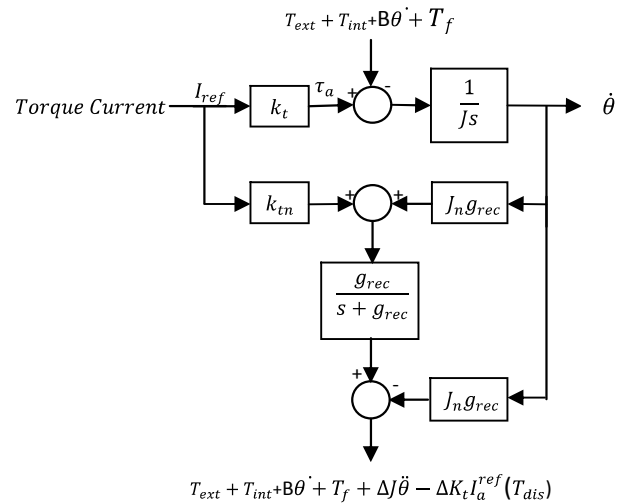


Fig.5. Estimation of Disturbance Torque using DOB

The disturbance torque is directly calculated from the acceleration of the motor and the real torque current. Here g_{rec} is the angular cutoff frequency of the low-pass filter.

Disturbance observer observes the disturbance force in the system without using force sensors. It is designed such a way that it is possible to estimate the disturbance from the observer.

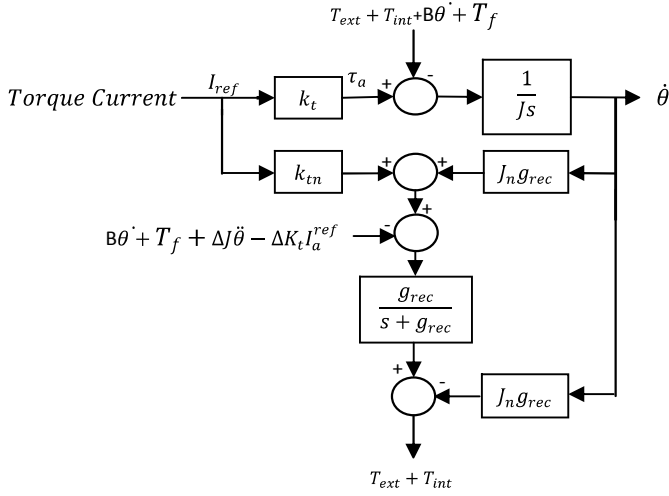


Fig. 6. Estimation of Reaction Torque using RTOB

Fig.6 shows the block diagram of the RTOB. Here, ΔJ is the self-inertia variation and ΔK_t is the self-motor constant variation. To estimate only the external torque, the torque exerted on the hardware setup due to friction component should be reduced. Therefore, in order to estimate only the external torque RTOB is used[8][9].

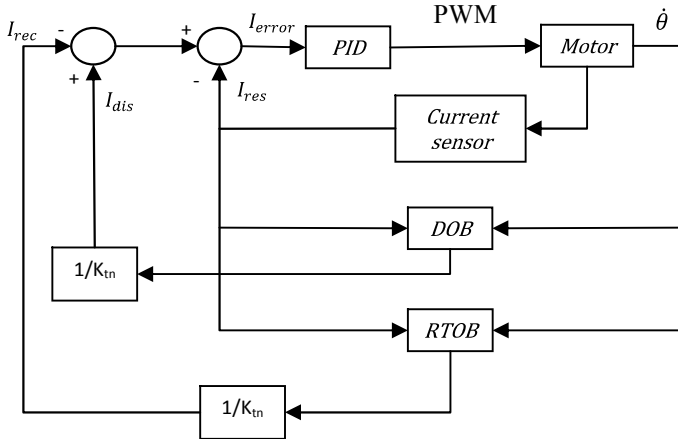


Fig. 7. Simplified version of DOB and RTOB Friction Compensator

Normally DOB and RTOB are very effective ways of finding disturbance torque values in current terms. Moreover disturbance torques can be calculated using those current values. But here both DOB and RTOB are included in given friction compensator (Fig.7). This system makes a perfect disturbance torque estimator.

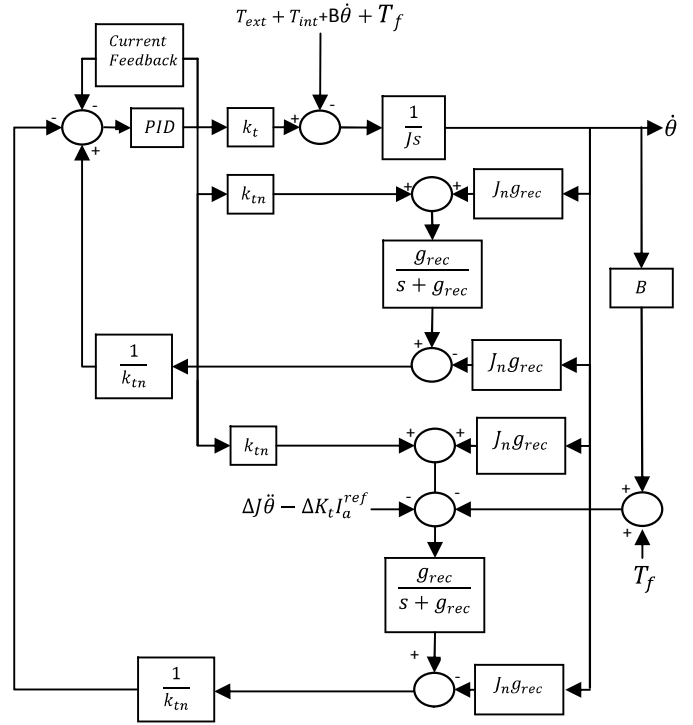


Fig. 8. Friction Compensator Block Diagram using DOB and RTOB Methods

VI. RESULTS

A. Stability Analysis

This paper analyzes the stability of the friction compensator system for Fig.8, as a basic control system of motion control. In order to do so, the simplified Transfer function (13) has been obtained in-between, the input is to be T_{dis} and output is to be $\dot{\theta}$ while the parameters as to be in table I.

TABLE I. THEORETICAL PARAMETERS

Parameters	Symbol	Value	Units
Nominal Motor inertia	J_n	0.000091	kgm ²
Motor inertia	J	0.000091	kgm ²
Nominal Torque coefficient	K_{in}	0.134	Nm/A
Torque coefficient	K_t	0.134	Nm/A
Proportional constant	K_p	0.001	rad/s
Integral constant	K_i	0.0001	rad/s
Derivative constant	K_d	0.00001	rad/s
Cut-off frequency of low pass filter	g_{rec}	200.0	Hz
Viscous friction coefficient	B	0.0000021	Nm/rpm
Static friction	T_f	0.0043119	Nm

$$\frac{-1.099}{s+1.396 \times 10^{-21}} \quad (13)$$

Suggested method gives a simplified transfer function and for stability analyzing purpose, the root locus method is used and it gives us only one root which is -1.396×10^{-21} root locus plot is given as in Fig.9. Apparently, it doesn't give oscillation part but only the exponential part. Since it's moving along the x axis towards positive infinity system is stable.

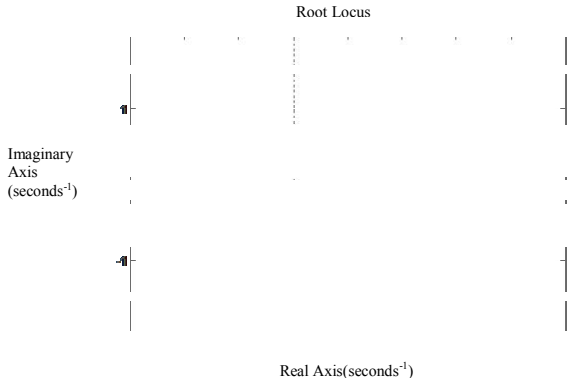


Fig. 9. Root Locus Plot for the Control Block Diagram

B. Simulation Results

The stability of the system was simulated for the Fig.8 under various types of disturbances. Initially a pulsive external torque has been applied to the rotor without friction compensator. It hardly rotates one revolution. Theoretically when an external pulse is applied to the rotor, it can be considered as T_{ext} and rotor will accelerate to a certain speed and continue to rotate in the same speed since the T_{dis} has been removed from the motor using the friction compensator. It is show in Fig.10 and Fig.11

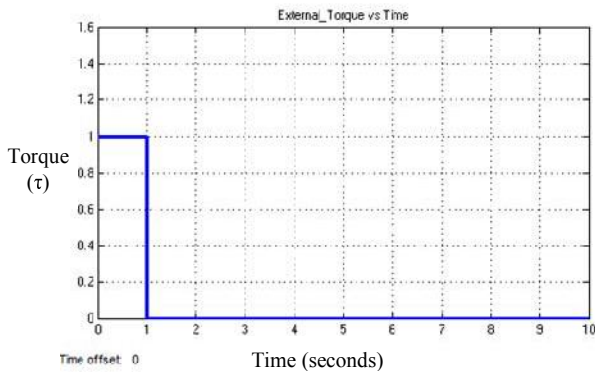


Fig. 10. External torque variation 1(A Pulse)

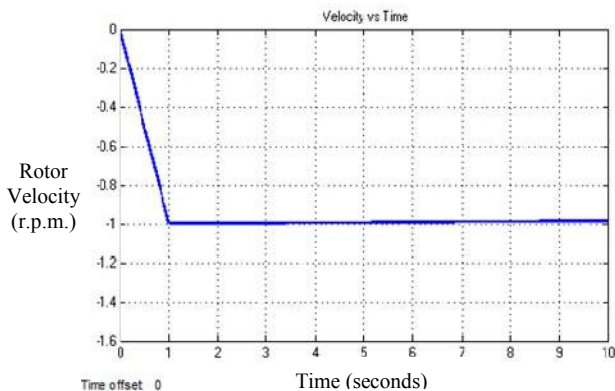


Fig.11. Motor Response 1(Rotating Speed Variation)

In the next scenario, two way pulses have been applied to the DC motor rotor manipulator. First clockwise external torque pulse has been applied and as in the previous scenario. It starts to accelerate to a certain speed and will rotate in the constant speed afterwards. And after few seconds the direction of the pulse has been reversed. And the equivalent torque is same as initial pulse. So theoretically rotor should decelerate and speed should become zero. It is drawn in Fig. 12 and Fig. 13 respectively.

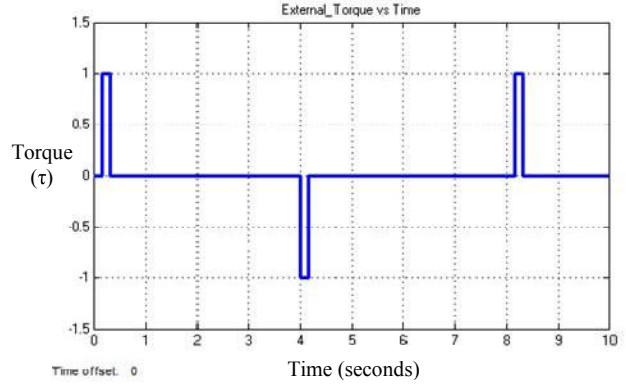


Fig.12. External Torque variation 2 (Two way Pulse Series)

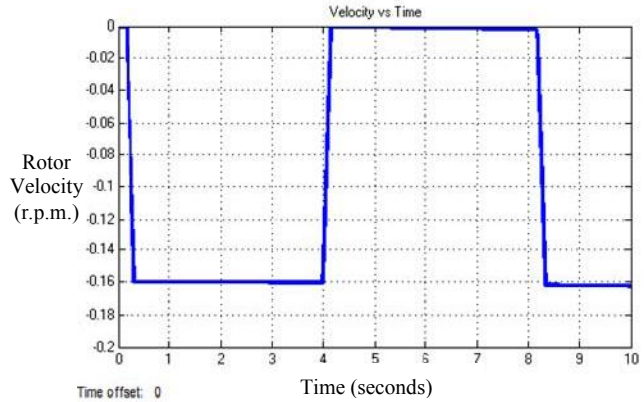


Fig.13. Motor Response 2(Rotating Speed Variation)

C. Practical Results

Practical scenario for the above simulation is carried out using the parameters shown in table I and it is shown in Fig.14. Here using a practical hardware setup actual results can be obtained. Fig.14 shows the response to single external torque pulse. As soon as the pulse is applied rotor starts to accelerate to a certain speed. If the system is adjusted correctly, smooth out put can be seen here.

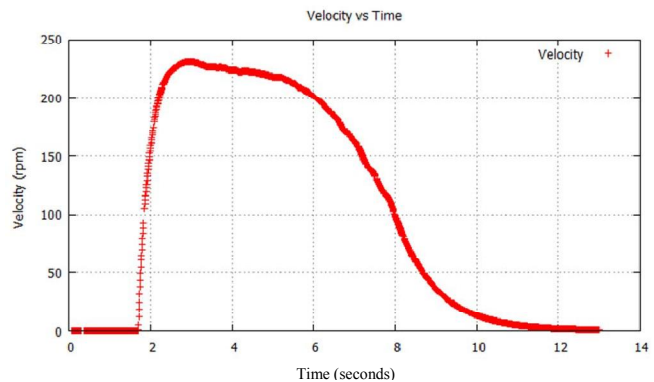


Fig.14. Motor Characteristics with friction Compensator-Single External Torque Pulse (Both F and $B\theta$ have been compensated)

VII. CONCLUSION

In this paper authors proposed a method for friction compensation that can be applied to brushed or brushless DC motors. Theoretically, this method can be simulated and will show the viable results which can be expected. System has been theoretically simulated under varies types of disturbances. Using solid practical results, the method has been evaluated possibility of using in practical applications. As far as practical scenario is concerned, there were problems aroused such as adjusting PID values. As for the results, the method is pretty much effective way of compensating the friction in a DC motor.

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REFERENCES

- [1] Xin Qian, Youyi Wang, “Motion Control of a Linear Brushless DCMotor Drive System with Nonlinear Friction Compensation” 2004 International Conference on Power System Technology - POWERCON 2004 Singapore vol.2, pp. 1276 - 1281, 21-24 November 2004.
- [2] Ivan Virgala, Peter Frankovský, Mária Kenderová, “Friction Effect Analysis of a DC Motor” Department of applied mechanics and mechatronics, Technical university of Košice / Faculty of mechanical engineering, Košice, Slovakia, <http://pubs.sciepub.com/ajme/1/1/1/>.
- [3] J. Annstrong-Helouvy B., Dupont P. and De Wit C., 1994. "A survey of models, analysis tools and compensation method for the control of machines with friction", *Automatica*, Vol. 30, No.7 (1994) pp. 1083-1138.
- [4] Ali, Y.S.E.; Noor, S.B.M.; Bashi, S.M.; Hassan, M.K.; "Microcontroller performance for DC motor speed control system," *Power Engineering Conference*, 2003. PECon 2003. *Proceedings. National*, vol., no., pp. 104- 109, 15-16 Dec. 2003.
- [5] Lawrence, D.A., “Stability and Transparency in Bilateral Teleoperation,” *IEEE Transaction on Robotics and Automation*, Vol.9, No.5, pp. 624-637, Oct. 1993.
Mizuochi, M.; Tsuji, T.; Ohnishi, K.; , "Improvement of disturbance suppression based on disturbance observer," *Advanced Motion Control*, 2006. 9th IEEE International Workshop on , vol., no., pp.229-234, 0-0-0.
- [6] Katsura, S.; Irie, K.; Ohishi, K.; , "Wideband Force Control by Position-Acceleration Integrated Disturbance Observer," *Industrial Electronics, IEEE Transactions on* , vol.55, no.4, pp.1699-1706, April 2008.
- [7] Hasala R. Senevirathne, A.M. Harsha S. Abeykoon, M. Branesh Pillai: "Disturbance Rejection Analysis of a Disturbance Observer based Velocity Controller," *ICIAIS'12*, Proceeding of the 6th IEEE Conference, Sep 27-29, 2012
- [8] K. Ohnishi, M. Shibata, T. Murakami : "Motion control for advanced mechatronics," *Mechatronics, IEEE/ASME Transactions on* , vol.1, no.1 pp.56-67, Mar 1996.
- [9] Dayarathna H.A.N.D., Prabuddha L.L.G., Ariyawansa K.L.D.N.J., Chinthaka M.K.C.D., Abeykoon, A.M.H.S., Branesh Pillai, M.: "Sensorless contact position estimation of a mobile robot in pushing motion," *ICCPCT, IEEE conference on Mar 20-21, 2013*.